# PELEONICS Number 2 1985

#### Editorial

One of the problems in building cave communication equipment has been the slow exchange of information, indeed obtaining information of any kind. Nowadays when everbody expects electronic equipment to halve in size and cost, and to double in performance every 5 (or is it 2?) years, it is discouraging to find so little information available, and many of the circuits dating back 20 or more years.

Clearly the Cave Communication Section through "Speleonics" can play a role in improving the spread of information, both circuits and ideas. One item in this issue is a <u>Bibliography</u>. To get things rolling I have listed all the articles of interest that I know. The idea is to get all the items into the NSS Library where they will be available for 10 cents a page, or swapped on a page for page basis for new material.

So please send me new listings so we can publish updates to the bibliography. (I. Drummond, 5619 Dalwood Way NW, Calgary, Alberta.T3A 1S6).

Consider sending Bill Torode, NSS Library, 1 Cave Avenue, Huntsville, Ala. 35810 a copy of any articles not already on the list, especially if you want material from the bibliography.

Other ways in which Speleonics can help communication is by suggestion of standards for equipment. Two items in this issue address that topic. One item is from the Eastern Region National Cave Rescue Commission concerning 12 V. power. If all units using 12 V power have the same connector and polarity, then on a rescue there is more chance of successful operation and less chance of damaged equipment. It made enough sense to me that I have changed my equipment to conform to the standard.

The second item concerns standardized frequencies. At present two of the more sucessful survey radios operate on frequencies of 3.500 kHz and 3.4956 kHz and are not fully compatible. As Frank Reid explains in his article, neither would work as well in the UK with 50 Hz power noise present. Perhaps a standard frequency could be agreed upon, so that in future radios of a particular type operate on a common frequency. This has been done, for example, with magnetic avalaunche beacons commonly carried by back-country skiers which operate at 2225 Hz world-wide.

Ian Drummond, Editor for Speleonics #2.

Editor for Issue #3 will be Frank Reid, PO Box 5283, Bloomington, IN, 47402

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#### Needed, A working Cave Radio.

I need to purchase a cave radio system as soon as possible.

Contact Dr. Nicholas C. Crawford
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# Report on Cave Communication and Electronics Section Meeting at Frankfort, Kentucky. June 24, 1985.

compiled from notes by J. Coward

The meeting started with a series of presentations.

- 1) Ray Cole talked about his "Organ Cave Radio". It is a narrow-band unit operating at 3.495 kHz for position-finding work only. The receiver has a novel commutating filter. Ray has submitted an article for Speleonics, complete with circuit diagrams.
- 2) <u>Ian Drummond</u> discussed the bibliography (published in this issue) and gave a little history of cave radio development, pointing out the split in direction between the US and the UK in the late 60's when the US went to narrow-band survey units and the UK went to higher frequency voice units. He also pointed out the US Bureau of Mines work which gives a much better design basis for radios.
- 3) <u>Julian Coward</u> outlined the ASS cave radio. It is a 2-way voice and tone system operating at 114.28 kHz carrier with 3 kHz upper side-band. A report with circuit diagrams has been published, but improvements have been made and an update will be available. Unlike other radio described at the meeting, the ASS radio antenna has a collapsible frame and stores in a tube for transport.
- 4) <u>Joe Hruska</u> described his 3.5 kHz survey unit, based on Frank Reid's design. Joe has not concerned himself over ultimate range, but emphasized robust equipment and simplicity.eg. his antenna is mounted in a bicycle tire and is flexible for transport into tight caves.
- 5) Frank Reid told how he had interconnected cave telephone systems with surface radios for cave rescue use. The first, a simple interconnection of microphone and speaker, worked but required an operator to switch Tx/Rx. The second method was a "Peter-porta", a commercial auto-repeater costing \$100. The third, completely sucessful and favoured system was the ICOM radio voice-operated transmit (VOX) switch.
- 6) Gene Harrison presented ideas on what was needed in a future radio design for cave rescue. a) 2-way speech plus emergency location. b) Range of 1000 ft. extendable to 1 mile (horizontally). c) Simple operation d) Standard frequency. e) Easy to build, circuit board with easy to obtain parts. f) Underground (1 cubic ft, (25 lb. g) Power 4-12 V DC.

#### Discussion

There was general concensus that the narrow-band 3.5 kHz units worked well for survey work.

There was clearly a demand for 2-way voice units. However there was no agreement on frequency or mode of transmission (AM, FM, or SSB). This lead in turn to a discussion of legal requirements. In the US no licence is required below 10 kHz; nor in a band 160-180 kHz for less than 1 watt of power.

One identified problem was lack of test proceedures for cave radios so it was very hard to compare relative performance before deciding what to build.

Good design proceedures now exist (Bureau of Mines work) but are not accesible to non-electrical engineers. There is a need for information on electrical noise-levels and ground conductivity measurments in caving areas.

Remote sensing of caves was briefly mentioned. Electromagnetic conductivity measurements (as opposed to current injection methods) have not been used by cavers. Also radar methods have been used to detect chambers in dolomite to 150 ft. depth, as well as to measure ice thickness on glaciers.

As a result of the discussion a "To do " list was started with names of volunteers to persue the topic.

2) 3)	FCC Regulations on Cave Radios (USA) DOC Regulations on Cave Radios (Canada) European Regulations on Cave Radios	Ray Cole Julian Coward
4)	Standardized tests on Cave Radios	Joe Hruska
5)	Theoretical Range calculations	Ian Drummond
	EM Conductivity Measurements	
	EM noise measurements	
8)	Adaptations of commercial equipment	
	to voice communication	

7) Remote sensing (Radar)

10) Design of a "state-of-the-art" survey unit at a frequency near 3.5 kHz.

Ian Drummond

Ray Cole and Frank Reid

Several people expressed interest in building radios for voice, but ideas were not concrete enough at the meeting to set-up groups to work on specific projects. Some of these people might be identified by the questionaire in this issue, which was also distributed at this meeting. See Speleonics 3# for results of the survey.

If anyone is interested on working on any of these projects, give the person a call, or if no-one is listed, sign-up by giving the new chairman (Ian Drummond) a call at (403)-288-4034.

An election was held at the lunch for the 85-86 period.

Elected Chairman Ian Drummond

Secretary Frank Reid Treasurer Joe Giddens

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Magnetic Moments, Number 2, by Ian Drummond

In this article I want to address the problem of design of loop antennas for transmission.

How big an antenna is needed for a given field, what gauge wire should be used, and how much power can an antenna handle?

In general a designer starts from one of two points. Either he can decide on the maximum size of antenna that can be physically handled in the cave or on the surface, or he starts knowing he must achieve a certain range. In either case he will want the greatest range for a given power.

The type of antenna considered here is the air-cored loop with a tuned secondary winding. I am assuming the designer is familiar with the concepts of transformers and resonant circuits, if not I would recommend reading an article such as The Amateur Radio Handbook, Chapter 2, on Electrical Laws and Circuits to gain an understanding of inductance, capacitance, and concepts such as "Q" of a resonant circuit.

The problem with using these basic concepts to design loop antennas, is that two effects, the skin effect, and the proximity effect, increase the apparent resistance of the wire to AC current over its DC resistance, and so degrade the performance of the antenna. These effects are treated in a very useful book "VLF Radio Engineering" by A D Watt, Pergamon Press, (1971), pages 90-101. I have incorporated the necessary formula into a Basic computer program which seems to work well for 3 very different sized loops, all working at 115.4 kHz, on which I have electrical measurements.

The AC resistance is calculated from the formula

 $R(AC) = R(DC) \times (1 + F + G(K + U))$  Ohms where

F = Skin Effect

6.K = Proximity effect due to nearby wires.

G.U = Proximity effect due to the magnetic field of the solenoid

These values are part of the output of the program and so allow the designer to estimate their importance in degrading the magnetic moment (NIA) of the loop.

CAVE RADIO ANTENNA DESIGN BY IAN DRUMMOND

VERSION 3 ; 85-07-02 SELECT ANTENNA SHAPE C=CIRCLE S=SQUARE P=PENTAGON H=HEXAGON? S RADIUS OR SPOKE LENGTH (METRES) 0.51 PERIMETER= 2.885 AREA= 0.5202 M^2 FREQUENCY (KHZ) =? 115.4 WIRE SIZE(8-32 AWG) SKIN EFFECT= 9.53E-03 X DC RESISTANCE **#TURNS IN SECONDARY** COIL? 70 LENGTH OF WIRE = 201.9 METRES MASS OF WIRE = 0.14 Kg FOR COIL ENTER # OF LAYERS? 1

WIRE SIZE. FOR MAGNET WIRE ENTER M , IF NOT ENTER DIA(CM)? M IS COIL CLOSE-WOUND (Y/N)? Y FORMULA FOR WIRE PROXIMITY EFFECT WILL ESTIMATE LOW COIL LENG 2.55 CM INDUCTANCE 1.10E-02 H FOR RESONANCE C= 1.73E-10 F NEARBY WIRE EFFECT= 9.18E-02 X DC RESIST SOLENOID EFFECT= 9.32E-02 X DC RESIST. ENTER CAPACITOR Q ( 200 IF VALUE IS UNKNOWN ) ? 200 CAPACITOR EQUIV. SERIES RESISTANCE = 39.9 OHMS CALCULATED Q= 87. Q(ACTUAL)? 85 COIL DC OHMS= 43.22 COIL AC OHMS= 91.50 COIL EQUIV OHMS= 93.82 POWER TO ANTENNA (WATTS)? 10 NIA= 1.19E+01 AMP.M2 AMPS(RMS) = 3.26E-01VOLTS(P-P) = 7.4E+03# TURNS IN PRIMARY ? 1 INPUT IMPED. TO ANTENNA=138.3 OHMS

Design of an antenna using this program becomes a question of trying various combinations of antenna size and shape, wire size, # of turns, etc. to find a combination that gives a good value of the magnetic moment (NIA) without exceeding voltage or current capabilities of the capacitors or wire.

In general for my own applications I have found all parameters except the input impedence to match well. The input impedence was too high by a factor of 2 or so. Frank Reid compared calculations with the actual values for some of his 3.5 kHz loops and found results within a factor of two also. So the program is not perfect, but I believe it is helpful in designing efficient antenna.

#### \*\*\*\*\*\*\*\*\*\*\*\*

All these examples are for a square antenna with a spoke length of 0.5 m. operating at a frequency of 115.4 kHz and with a single turn primary.

```
# Turns
            Wire(AW6)
                        NIA(10 watts)
                                         Peak Volts
              28
                           11.7
                                           7.4 kV
                                                       87
(This is the calculation for one of our field antennas and matches well.)
              28
                           14.5
                                           15 kV
                                                      113
(Doubling the turns only increases NIA by 24% but fries the 6 kV capacitors and the
bandwidth is too small)
                           13.9
                                           7.5 kV
                                                      115
(Same # turns but thicker wire increases NIA by 19% only and the bandwidth is still too
narrow)
 35
                           11.2
                                           3.5 kV
(Maybe this would have been a better choice than our original one, less work to make,
anyway).
```

Julian tried an empirical approach to antenna design. A square antenna with 1.1 m spokes, and a one turn primary.

4 16 21.1 0.48 kV 39 (Not bad, certainly a lot easier to make, but big for handling in the bush)

#### And so it goes.....

```
10 PRINT"CAVE RADIO"
                                170 INPUT "LENGTH (METRES)",RA 350 PRINT USING"AREA=##.#### M
20 PRINT"ANTENNA DESIGN"
                                                                ^2": AREA
40 PRINT"BY IAN DRUMMOND":PRI 180 IF (SHAPE = "C") THEN 230
                                                                360 ERAD=SQR(AREA/PI)
NT
                                190 IF (SHAPE = "S") THEN 260
                                                                380 INPUT "FREQUENCY(KHZ) = "; FR
60 PRINT"VERSION 3 ; 85-07-02 200 IF(SHAPE$="P")THEN 290
":PRINT
                                210 IF(SHAPE$="H")THEN 320
                                                               390 SKIN=1/(SQR(PI*23.2*PI*FRQ
70 PRINT"PROGRAM WORKS BEST":P 220 GOTO 100
                                                                *1000))
RINT*WITH SINGLE LAYER*:PRINT* 230 PERIM=2*PI*RAD
                                                                400 PRINT"WIRE SIZE(8-32 AW6)"
SPACED WIRE-COILS":PRINT"WILL 240 AREA=PI*RAD*RAD
                                                                : INPUT AWS
OVERESTIMATE": PRINT" OTHERWISE 250 60T0 340
                                                                410 IF AWG(8 OR AWG>32 OR (AWG
                                260 PERIM=SQR(2)*4*RAD
                                                                /2<>INT(AWB/2))THEN 400
80 PI=3.1416
                                270 AREA=(PERIM/4)^2
                                                               420 RESTORE
100 PRINT"SELECT ANTENNA SHAPE 280 GOTO 340
                                                                430 FOR I=1 TO (AWG/2-3)
                                                                440 READ DIA, RESDC: NEXT I
                                290 PERIM=SIN(PI/5)*10*RAD
                                                                450 DATA .326,.00206,.259,.003
110 PRINT" C=CIRCLE
                                300 AREA=1.72*(PERIM/5)^2
120 PRINT"S=SQUARE"
                                                                28,.205,.00522,.163,.00827,.12
                                310 GOTO 340
130 PRINT" P=PENTAGON"
                                320 PERIM=6*RAD
                                                                9,.0132,.102,.021,.0813,.0331,
140 PRINT" H=HEXABON":
                                330 AREA=2.598*RAD^2
                                                                .0643,.0531,.0511,.0843,.0404,
150 INPUT SHAPE$
                                340 PRINT USING"PERIMETER=##.# .135,.032,.214,.0254,.341,.020
160 PRINT "RADIUS OR SPOKE"
                                ##"; PERIM
                                                               3,.531
```

460 IF(DIA>(400\*SKIN))THEN 490 470 RESAC=RESDC \* (1+(((DIA/(40 0\*SKIN))^2)^2)/3));GOTO 520 490 IF(DIA>1000\*SKIN)THEN 510 500 RESAC=RESDC\*(.55+.00195\*DI A/SKIN):GOTO 520 510 RESAC=RESDC\*DIA/(400\*SKIN) 520 PRINT"SKIN EFFECT=" 540 PRINT USING"##.##^^^ X DC RESISTANCE"; RESAC/RESDC-1 560 PRINT"#TURNS IN SECONDARY" 570 INPUT"COIL"; N 572 RESTORE 574: FOR I = 1 TO (A WS/2-3) 573 READ MASS: NEXT I 574 DATA .0745,.0469,.0295,.01 85,.01165,.00734,.00462,.00289 ,.00182,.00114,.000716,.000452 ,.000289 575 PRINT USING"LENGTH OF WIRE =####.# METRES"; N\*PERIM 576 PRINT USING "MASS OF WIRE = ##.## Kg";N\*PERIM\*MASS 580 PRINT"FOR COIL ENTER" 590 INPUT"# OF LAYERS"; NL: PRIN Τ 600 CLS:PRINT"WIRE SIZE." 610 PRINT" FOR MAGNET WIRE" 620 PRINT" ENTER M , IF NOT" 630 INPUT" ENTER DIA(CH)";DIA\$ 640 IF(DIA\$="M")THEN 670 650 WDIA=VAL(DIA\$) 660 GOTO 680

670 WDIA=DIA\*(1+AW6/200) 680 PRINT"IS COIL CLOSE-WOUND" 690 INPUT"(Y/N)"; A\$ 700 IF A\$="Y" THEN 730 : IF A\$< >"N"THEN 680 710 INPUT"# TURNS/CH";TURNS:IF TURNS>1/WDIA GOTO 710 720 L=N\*.01/(TURNS\*NL):60T0 77 730 IF WDIA/DIA>1.7 GOTO 760 740 PRINT"FORMULA FOR WIRE":PR INT"PROXIMITY EFFECT" 750 PRINT"WILL ESTINATE LOW" 760 L=.01\*N\*WDIA/NL 770 PRINT USING "COIL LENG #### .## CM";L\*100 780 MU=1 790 60SUB 2160 800 PRINT"INDUCTANCE" 810 PRINT USING"##.##^^^ H";I ND\*.000001 830 CAP=IND/((IND\*2\*PI\*FRQ)^2)

840 PRINT\*FOR RESONANCE\* 850 PRINT USING"C=##.##^^^ F" ; CAP 855 REM CALC WIRE PROXIMITY EF FECT 860 IF DIA\*.01/SKIN>2 THEN 870 865 GBAR=(DIA\*.0025/SKIN)^4:GO TO 890 870 IF DIA\*.01/SKIN>4 THEN 880 875 6BAR=.0015\*DIA/SKIN-.238:6 DTO 890 880 GBAR=.12\*(.01\*DIA/SKIN-1) 890 REM NEARBY WIRE EFFECT ONL Y ACCURATE IF (CENTRE TO CENTR E)/DIA>1.7 900 IF N<100 THEN 910 905 K=3.3:GOTO 920 910 K=1.58\*L06(N)-.194\*(L06(N) ^2) 920 PRINT"NEARBY WIRE EFFECT=" :PRINT USING "##.##^^^ X DC R ESIST"; K\*6BAR 930 REM SOLENOID EFFECT 935 U=3.19+1.1\*L/ERAD-.065\*((L /ERAD)^2) 940 PRINT"SOLENOID EFFECT=":PR INT USING "##.##^^^ X DC RESI ST.";U\*6BAR 945 PRINT "ENTER CAPACITOR Q": PRINT" ( 200 IF VALUE IS UNKNOW N )" 947 INPUT CAPQ 949 RCAP=1/(2000\*PI\*FRQ\*CAP\*CA 951 PRINT USING "CAPACITOR EQU IV. SERIES RESISTANCE =###.# 0 HMS "; RCAP 960 PRINT"CALCULATED Q"; 970 R=N\*PERIN\*RESDC\*((RESAC/RE SDC)+K\*GBAR+U\*GBAR)+RCAP:Q=.00 2\*PI\*FRQ\*IND/R 980 PRINT USING "=####"; Q 990 INPUT"Q(ACTUAL)";QA 1000 PRINT USING\*COIL DC OHMS= ###.##"; N\*RESDC\*PERIM

1020 PRINT USING COIL EQUIV OH MS=###.##"; REFF 1040 PRINT"POWER TO ANTENNA": I NPUT" (WATTS) "; WATTS 1050 AMPS=SQR(WATTS/REFF) 1060 VPP=.00566\*AMPS\*PI\*FRQ\*IN 1070 NIA=N\*AMPS\*AREA

1010 PRINT USING COIL AC OHMS=

1015 REFF=(.002\*PI\*FRQ\*IND/QA)

###.##";R

1080 PRINT USING "NIA=##.##^^^^ AMP.M2"; NIA 1090 PRINT USING AMPS (RMS) = ##. \*\*^^^^#; AMPS 1100 PRINT USING "VOLTS (P-P) = ## . #^^^^\*; VPP

1120 SECIMP=QA\*.002\*PI\*FRQ\*IND 1130 PRINT"# TURNS IN PRIMARY" :INPUT PN 1140 PRINT"INPUT IMPED. TO":PR INT USING "ANTENNA=###.# OHMS"; SECIMP\*((PN/N)^2) 1160 PRINT"TO RECALC, ENTER #\* 1170 PRINT"1-PRI.TURNS 4-POWER 1180 PRINT"2-SEC.TURNS 5-AW6" 1190 PRINT"3-ACTUAL Q 6-FREQ" 1192 PRINT" 7-SIZE" 1200 A = INKEY =: IF A = \* \* THEN 1 200 ELSE 1210 1210 CLS: A%=VAL(A\$): ON A% 60TO 1130,560,990,1040,400,380,100 2000 GOTO 2310 2160 REM CALC INDUCTANCE 2170 RATIO=2\*ERAD/L 2180 IF(RATIO<.01 OR RATIO>100 0) THEN 2250 2190 IF(RATIO>1)THEN 2220 2200 FBAR=RATIO+(.025-8.99999 E-03\*RATIO) 2210 GOTO 2230 2220 FBAR=.016+.0156\*LD6(RATIO 2230 IND=MU\*78.8\*FBAR\*N\*N\*ERAD

2240 RETURN 2250 CLS:PRINT\*COIL DIMENSIONS ":PRINT"OUTSIDE RANGE" 2260 GOTO 2310 2310 END

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## Thoughts towards designing an "International" cave radio: Choosing an operating frequency.

by Frank Reid

A primary goal of the NSS Electronics Section is designing high-performance, easily built cave radios. Of the few sets of equipment that now exist, most are on different frequencies. Cave radios on common frequencies (I don't want to use the word "standard" at this time) would be advantageous in cave rescue and other inter-group efforts, and for performance tests. This article applies to CW radios operating at mid range audio frequencies; cave radios which transmit voice will be covered elsewhere.

Nevin Davis (1) has shown that frequencies near 3.5 kHz are optimum for magnetic induction cave radio. Power-line interference is a major problem for radios having carrier frequencies in the audio range. Harmonics of the 60 Hz power-line frequency are strongest below 600 Hz, but extend to higher frequencies. The strongest 60 Hz harmonic is 180 Hz, and is probably caused by a 3-phase effect.

Power-line interference can be limited by choosing an operating frequency that falls between power-line harmonics, and using a receiver bandwidth narrow enough to reject the adjacent harmonics. Europe uses 50 Hz power; a cave radio frequency which falls in between the harmonics of both 50 and 60 Hz would be workable in any country. I have operated a 3500 Hz cave radio for more than 12 years (3); 50 Hz harmonics are not a problem in America. Harmonics of 60 Hz probably would not be detected in Europe.

Harmonics of 50 and 60 Hz coincide every  $300\ \text{Hz}$ , and the pattern of seperation repeats every  $300\ \text{Hz}$ .

50 100 150 200 250 300 350 400 450 500 550 <u>600</u> 650 700 60 120 180 240 <u>300</u> 360 420 480 540 <u>600</u> 660 720

We can never get more than 25 Hz away from a harmonic of 50 Hz, nor 30 Hz away from a 60 Hz harmonic. The largest "windows" between harmonics are 50 Hz wide, and occur either side of multiples of 300 Hz. Assuming that we wish to avoid any harmonic by at least 20 Hz, the most desirable frequency for an "international " cave radio in the 3.5 kHz region would be 20 to 30 Hz on either side of a multiple of 300 Hz, eg. 2700, 3000, 3300, 3600, or 3900 Hz.

Cave radios having very narrow receiving bandwidths require quartz crystal oscillators for frequency stability. We would like to use commonly available crystals, and use the same crystal frequency in both transmitter and receiver. (Receivers use local oscillators to drive mixers and for clocking digital filters and correlators). We would also prefer that the crystal frequency divisor be an even number, so that the divider output is a symetrical square wave.

The 3.579545 MHz colour-burst crystals used in colour televisions are inexpensive and more readily available than any other crystals. (Are these available in Europe? Editor). 3579.5 Hz (the colour TV frequency divided by 1000 ) satisfies all the above criteria, being 20.5 Hz away from 3600 Hz, the nearest harmonic of 50 or 60 Hz.

3276.8 Hz, made by dividing an electronic wristwatch oscillator frequency by 10, also fulfills the requirements for harmonic avoidance, easy parts availability and an even-numbered divisor. It is 23.2 Hz from 3300 Hz.

Many other candidates exist.

```
32768/12 = 2730.67
1 \text{ MHz}/(21 \times 16) = 2976.19
4 \text{ MHz}/(\$67 \times 16) = 2427.18
4 \text{ MHz}/(\$88 \times 8) = 2673.80
4 \text{ MHz}/(\$67 \times 32) = 3023.26
4 \text{ MHz}/(\$88 \times 8) = 2673.80
4 \text{ MHz}/(\$67 \times 16) = 3924.94
4 \text{ MHz}/(\$11 \times \$12) = 3023.26
4 \text{ MHz}/(\$11 \times \$12) = 3023.30
4 \text{ MHz}/(\$11 \times \$12) = 3023.30
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```

Where \$ = hexadecimal number instead of a three digit decimal number

It is easy to divide a frequency by any integer, using programmable up/down counters, eg. 74192 or 74193 TTL chips or their CMOS equivalents.(4) Converting decimal divisors to hexadecimal and using full 4-bit counters instead of decade counters sometimes saves chips, eg. decimal numbers from 100 - 255 become 2-place numbers when converted to hexadecimal.

The simplest frequency divider is a cascade of flipflops, which divides frequencies by powers of two. Filter-clocking and local oscillator frequencies are easily obtained by tapping a straight divider chain (available on a single chip). Cave radio experimenter Ray Cole has had excellent results using the TV crystal of 3.579545 MHz and such a chain to provide frequencies of 3.49565 and 6.9913 kHz combined via an "AND" gate for a 25% duty cycle transmit pulse at 3.49565 kHz, and frequencies of  $873.9 \times 1$ , x 2, x 4, x 8, x 16 to clock his commutating filter. Unfortunately while his transmitter frequency of 3495.65 Hz fits well between 60 Hz harmonics it is only 4 Hz away from a 50 Hz harmonic.

The US Bureau of Mines tested 630, 1050, 1950 and 3030 Hz for mine-rescue radio location (2). Results were slightly better at lower frequencies. All these frequencies are half-way between 60 Hz harmonics but only 630 and 3030 Hz avoid 50 Hz harmonics by the formula above.

Ideally, local oscillator frequencies and mixer products should also be in between power-line harmonics. Intermodulation could cause interference if the input signal to the mixer contains powerline harmonics near the local oscillator frequency.

#### References

- (1) Davis, Nevin W.Optimum frequencies for underground communication. NSS Bulletin vol 32, #1, (1970)
- (2) Durkin, John. Performance evaluation of electromagnetic techniques for the location of trapped miners, US Bureau of Mines, Report of Investigation, RI 8711,(1982).
- (3) Reid, Frank Caveman Radio. 73 Magazine, Feb, (1984)
- (4) Reid, F and Honeycutt, R K. A digital clock for sidereal time. Sky and Telescope, July, p59, 1976.

#### NEEDED

A program for an IBM-PC for drawing electronic circuits.
Preferably cheap and with templates if necessary.
Help your editors give what we all want.....

SHORT ARTICLES AND LONG CIRCUITS not LONG ARTICLES AND SHORT CIRCUITS.

Info by letter or phone to Ian Drummond, (403)-288-4034 or see questionaire for address.

### Cave Communication Questionaire

Address	Are you representing a group? Y/N  If "yes", who are they, and how many people  are there?  Group Name  # of people
Do you alre If "yes",	please give a brief description
loan equipm	Is it satisfactory in performance? Y/N
Which techn	iques are you interested in? Telephone Seismic Radio Other
What do you	want to use the equipment for <u>primarily</u> ?  Survey work Emergency location  Voice communication for Rescue  Exploration  Underwater  Data transfer Remote sensing
What depths	Other, please describe
What length trip?	of time do you need the equipment to operate on an underground hours days
Are you int	erested in Field testing of equipment?  Construction of established designs?  Construction of experimental designs?  Work on the theory for a better design basis?  Organization of this NSS Section?  Other?
the NSS Lib	any books or articles which you would like to add to the bibliography or to rary on this topic?
If you are day for con	a Radio Amateur, please include call-sign, preferred frequency, and time of tacts.
Please send	the completed questionaire to: I. Drummond

5619, Dalwood Way NW Calgary, Alberta, Canada

T3A 156

Any additional correspondence welcome.





# APPALACHIAN SEARCH AND RESCUE CONFERENCE, INC. P.O. BOX 440 NEWCOMB STATION CHARLOTTESVILLE, VIRGINIA 22901

COMMUNICATIONS STANDARD #1.0
12 VOLT D.C. POWER (JONES PLUGS")

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REVISION O P1/2

Purpose: STANDARDIZE AND PROMOTE INTERCHANGABILITY OF

12 VOLT D.C. AUTOMOTIVE-TYPE POWERED DEVICES AMONG

EMERGENCY PERSONNEL AND ORGANIZATIONS

ELECTRICAL: "12 VOLTS" RANGING FROM 11 TO 15, BUT TYPICALLY

13.8 VOLTS DIRECT CURRENT, USUALLY FROM AUTOMOBILES

OR BATTERIES, AT UP TO 7.5 AMPS (10 AMPS MAXIMUM).

STANDARD: 2-PIN "JONES PLUG" AS ILLUSTRATED BELOW.

SOURCE AND NUMBERS: CORD MALE CORD FEMALE CHASSIS FEMALE

RADIO SHACK -> 274-201 274-202 - 274-203

TRW CINCH -> P-302-CCT S-302-CCT P-302-AB S-302-AB

(TRW VERSIONS RECOMMENDED: METAL SHELLS, MORE VARIETIES AVAILABLE)

CONNECTION AND ASSEMBLY: SOLDER(!) AND INSULATE CONNECTIONS. (\* = TYPICAL)

MALE (PINS) FOR LOADS (RAPIOS, LITES...)

NARROW FOR "HOT", +12V, REDWIRES

FEMALG (HOLES) FOR SOURCES (BATTERY...)

WIDE FOR "GROUND, -12V, BLACKWIRES

SAFETY: ALWAYS FUZE-PROTECT WITH RATING APPROPRIATE TO 1) SPECIFIC DEVICE/LOAD,

2)
SPECIFIC SOURCE/BATTERY, OR WIRE SIZE, WHICHEVER IS LEAST, OR FUZE

CONNECTOR AT IDAMP (MAX 15).

WIRE SIZES/MAX AMPS: AWGIG @ 7A; "14@15A; "12@2DA; "10@3DA

USE HIGH TEMPERATURE AUTOMOTIVE-TYPE WIRE, ESPECIALLY IN CARS OR ROUGH SERVICE.

LIGHT-DUTY CORDS AND ADAPTORS MIGHT USE LAMP (ZIP) CORD.

-COORDINATION	-COMMUNICATION	-TRAINING

@ INOVSYGLHARRISON

Revision 0 P2/2

#### 2-PIN JONES PLUG:

( = TYPICAL)











